Metaheuristic Optimization Techniques for an Electromagnetic Multilayer Radome Design

Trung Kien Nguyen¹ · In-Gon Lee¹ · Obum Kwon² · Yoon-Jae Kim² · Ic-Pyo Hong^{1,*}

Abstract

In this study, an effective method for designing an electromagnetic multilayer radome is introduced. This method is achieved by using ant colony optimization for a continuous domain in the transmission coefficient maximization with stability for a wide angle of incidence in both perpendicular and parallel polarizations in specific X- and Ku-bands. To obtain the optimized parameter for a C-sandwich radome, particle swarm optimization algorithm is operated to give a clear comparison on the effectiveness of ant colony optimization for a continuous domain. The qualification of an optimized multilayer radome is also compared with an effective solid radome type in transmitted power stability and presented in this research.

Key Words: Ant Colony Optimization, Electromagnetic Multilayer Radome Design, Metaheuristic Optimization Algorithm, Particle Swarm Optimization, Transmission Coefficient Maximization.

I. INTRODUCTION

The problem of an electromagnetic multilayer dielectric design optimization for a frequency band and a desired range of incident angles has been introduced in recent years using such metaheuristic optimization algorithms as the genetic algorithm [1, 2], particle swarm optimization (PSO) [3, 4], or a hybrid algorithm that combines ant colony optimization (ACO) with the microgenetic algorithm [5]. In this study, an effective algorithm presented by Socha and Dorigo [6] in 2008 called ant colony optimization for a continuous domain (ACO_R) is applied to design a C-sandwich radome [7] with an applicable range of incident angle (0°–70°) in both the transverse electric (TE) and transverse magnetic (TM) modes for the X- and Ku-bands. To validate the ACO_R algorithm in a multilayer radome

design optimization, the performance of the transmission coefficient characteristics of the C-sandwich radome design optimized by ACO_R has been compared with a conventional analysis method (i.e., a simpler and lower-cost analysis) that approximates multilayer radomes by a solid radome with an effective medium approximation (EMA) [8, 9]. Then, the performance of ACO_R is compared with a general optimization algorithm used in electromagnetic characteristic design (i.e., PSO). This study is organized as follows: Section II is an overview of the ACO_R algorithm, which uses a boundary value method to evaluate the transmission coefficient of a multilayer radome. Section III examines the fitness function and compares the optimization results obtained by ACO_R and PSO. The simulation results and the transmitted power stability are compared with an effective solid radome, and the results are discussed. The conclusion is

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provided in Section IV.

II. METHODS

1. Ant Colony Optimization for a Continuous Domain

Originally, ACO was introduced in 1992 by Dorigo [10], and it has been used to solve many combinational optimization problems consisting of a set of discrete decision variables. The idea of applying ACO in solving the continuous optimization problem was presented in 2008 by Socha and Dorigo [6]. The flowchart of the ACO_R algorithm is illustrated in Fig. 1.

In ACO_R, the solutions are kept and sorted in a solution archive (Fig. 2), in which the associated weight of solution l defined in Eq. (1) represents the strength of the solution in producing new solutions.

$$\omega(l) = \frac{1}{qk\sqrt{2\pi}} e^{-\frac{(l-1)^2}{2q^2k^2}}. \quad (1)$$

To generate new solutions, a kernel is selected by probability that is computed for each group as in Eq. (2).

$$p(l) = \frac{\omega(l)}{\sum_{j=1}^k \omega(j)}. \quad (2)$$

Roulette wheel selection [11] is applied to select a solution kernel. Each probability $p(l)$ is presented as a proportion of the wheel (Fig. 3), and a random selection process is made similarly to rotate the roulette wheel. An m new random number

Fig. 1. Flowchart of ACO_R algorithm.

s_1^1	s_1^2	$\dots$	s_1^i	$\dots$	s_1^n	$\omega(1)$	$f(1)$
s_2^1	s_2^2	$\dots$	s_2^i	$\dots$	s_2^n	$\omega(2)$	$f(2)$
$\vdots$	$\vdots$	$\ddots$	$\vdots$	$\ddots$	$\vdots$	$\vdots$	$\vdots$
s_l^1	s_l^2	$\dots$	s_l^i	$\dots$	s_l^n	$\omega(l)$	$f(l)$
$\vdots$	$\vdots$	$\ddots$	$\vdots$	$\ddots$	$\vdots$	$\vdots$	$\vdots$
s_k^1	s_k^2	$\dots$	s_k^i	$\dots$	s_k^n	$\omega(k)$	$f(k)$

Fig. 2. Solution archive of the ACO_R algorithm with s_l^i as the value of parameter i in group l and the objective value $f(l)$ with associated weight $\omega(l)$.

Fig. 3. Roulette wheel with a probability of solution $p(l)$ as the proportion of the wheel.

group according to the parameterized normal distribution is used with mean (μ_l^i) and standard deviation (σ_l^i) for ant i in group l .

$$\mu_l^i = s_l^i. \quad (3)$$

$$\sigma_l^i = \xi \sum_{j=1}^k \frac{|s_j^i - s_l^i|}{k-1}. \quad (4)$$

$\xi > 0$ is a constant parameter that is similar to the pheromone evaporation rate in ACO [10]. Then, m new solutions are added to the solution archive and re-ordered. Non-fit solutions are removed, and only the k best ones are kept after each iteration.

2. Transmission Coefficient Characteristic of the C-Sandwich Radome

A C-sandwich radome, which is a multilayer radome (style d) [7], is considered in this work. The sandwich wall has five layers that were developed to cover an antenna array operating in a wide-band frequency range. The C-sandwich construction consists of three skins that are interleaved by two cores. Typically, the relative permittivity of skin is greater than that of the core. To guarantee that the input and output of the propagation wave

are equal, a C-sandwich wall is designed with the first and last skins having the same thickness, and the widths of the two cores are the same as well. To fabricate a radome, the thickness of each layer is designed as the multiple of plies. The propagation wave on the radome with layer i has thickness d_i , relative permittivity ε_i , loss tangent $\tan\delta_i$, and refractive index n_i . The F_j ($j \in [1,7]$)/ B_k ($k \in [1,6]$) are the forward and backward propagation waves, respectively. The forward and backward fields at the first left interface are related to Eq. (5).

$$\begin{bmatrix} F_1 \\ B_1 \end{bmatrix} = \prod_{i=1}^5 \frac{1}{\tau_i \tau_6} \begin{bmatrix} e^{jk_i d_i} & r_i * e^{-jk_i d_i} \\ r_i * e^{jk_i d_i} & e^{-jk_i d_i} \end{bmatrix} \times \begin{bmatrix} 1 & r_6 \\ r_6 & 1 \end{bmatrix} \begin{bmatrix} F_7 \\ 0 \end{bmatrix} \quad (5)$$

$\tau_i = 1 + r_i$ and r_i is the intrinsic reflection coefficient of each interface and is calculated by the Fresnel equation. If one uses the formula in Eq. (5) and simplifies it to Eq. (6), the transmission coefficient can be obtained by Eq. (7).

$$\begin{bmatrix} F_1 \\ B_1 \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} F_7 \\ 0 \end{bmatrix}. \quad (6)$$

$$T = \frac{F_7}{F_1} = \frac{1}{A_{11}}. \quad (7)$$

III. RESULTS AND DISCUSSION

1. Fitness Function

The main objective of this research is to maximize the transmission coefficient of a multilayer structure radome, which is calculated by Eq. (7) in a specific frequency band with various angles of incidence in both the TE and TM modes. To ensure the stability of the transmission coefficient, the following objective function is considered for optimization:

$$\text{maximize } F = \{\min[T(f, \theta)]\}^2 * 100, \quad (8)$$

where $T(f, \theta)$ is the transmission coefficient at frequency f and incident angle θ .

2. Optimization Design Result

The optimized thickness design is compared with an effective solid radome type with a half-wavelength thickness. The transmitted power is 80% greater in both polarizations, and the wide angle of incidence is the target of optimization. The design of the multilayer radome structure for optimization is shown in Table 1.

According to the EMA method, a multilayer radome can be approximated by its effective material property. The effective relative permittivity (ε_{eff}) of the radome after optimization is calculated by Eq. (9), where n is the total number of layers, and d_i , ε_i are the thickness and permittivity of layer i , respectively. By using a half-wavelength equation to assume the thickness of the effective medium, Eq. (10) can be used to compute

Table 1. C-sandwich radome layer properties

Layer	Permittivity	Loss tangent	Thickness (mm)	Step/ply (mm)
Skin 1	4.4	0.016	0.48–2.4	0.24
Core 1	1.1	0.001	2–9	0.1
Skin 2	4.4	0.016	2.4–9.6	0.24
Core 2	1.1	0.001	2–9	0.1
Skin 3	4.4	0.016	0.48–2.4	0.24

the thickness of the effective half-wavelength solid-type radome at a resonant frequency of 10 GHz for the X-band and 15 GHz for the Ku-band with an angle of incidence $\theta = 70^\circ$.

$$\varepsilon_{eff} * \sum_{i=1}^n d_i = \sum_{i=1}^n \varepsilon_i * d_i. \quad (9)$$

$$d_{eff} = \frac{\lambda}{2 * \sqrt{\varepsilon_{eff} - \sin(\theta)^2}}. \quad (10)$$

The first optimized radome of this study is the design for the X-band (8–12 GHz) frequency. According to a comparative analysis with 30 execution times (Table 2), ACO_R requires less time (56.69 seconds) to reach the convergence than PSO (70.18 seconds). Fig. 4 presents the comparison of transmitted power stability between the optimized multilayer radome and the effective solid radome with the stability of a multilayer radome in various angles, with both TE and TM modes being under 10% change. The transmission coefficient characteristic of the C-sandwich radome after optimization is presented in Fig. 5 in the X-band.

To give a more effective decision, the same design properties of the multilayer radome are applied in the Ku-band (12–18 GHz). Table 3 shows the same result to demonstrate that ACO_R has better performance than PSO. Fig. 6 illustrates that the stability in the transmitted power of the optimized multilayer structure is less than 10%, which is outstanding compared with that of the effective solid radome. Fig. 7 shows the transmission coefficient of the optimized radome for the TE and TM modes in the Ku-band.

Table 2. Comparison of the 30 execution times between ACO_R and PSO for the X-band frequency

Detail	ACO _R	PSO
Skin 1 (mm)	0.72	0.72
Core 1 (mm)	5.0	5.0
Skin 2 (mm)	7.68	7.68
Core 2 (mm)	5.0	5.0
Skin 3 (mm)	0.72	0.72
Average time (s)	56.69	70.18
Convergent iteration	11	24

Fig. 4. Stability of the transmitted power of an optimized multilayer structure with an effective solid-type radome in the X-band.

Fig. 5. Optimized transmission coefficient of the multilayer structure for the X-band.

According to the computed result, ACO_R gives a promising value in optimizing multi-objective problems applied in the electromagnetic multilayer radome design.

IV. CONCLUSION

A comparative study of two metaheuristic optimization algorithms, PSO and ACO_R for an electromagnetic multilayer ra-

Table 3. Comparison of the 30 execution times between ACO_R and PSO for the Ku-band frequency

Detail	ACO_R	PSO
Skin 1 (mm)	0.48	0.48
Core 1 (mm)	3.3	3.3
Skin 2 (mm)	5.04	5.04
Core 2 (mm)	3.3	3.3
Skin 3 (mm)	0.48	0.48
Average time (s)	64.13	71.86
Convergent iteration	12	24

Fig. 6. Stability of the transmitted power of the optimized multilayer structure with an effective solid-type radome in the Ku-band.

Fig. 7. Optimized transmission coefficient of the multilayer structure for the Ku-band.

dome design with various angles of incidence (0° – 70°) in the perpendicular and parallel polarizations for the X- and Ku-bands was presented. The simulation results obtained by ACO_R were compared with those of a trusted optimization, PSO. The results strongly confirm that ACO_R can be useful in the electromagnetic multilayer radome characteristic optimization. The stability for the transmission coefficient of the optimized structure is also guaranteed by ACO_R . In future works, the ACO_R algorithm will be improved to solve more complex frequency-selective surface design optimization problems such as the 3D frequency-selective surface screen design.

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