I. INTRODUCTION

Compared with the conventional wired cable-connected endoscopies, capsule endoscopy provides better diagnoses and treatments throughout the entire gastrointestinal (GI) tract in a minimally invasive way, thus causing less discomfort and less pain to patients [1]. Recent wireless capsule endoscopies have been developed to capture higher-quality wide-view-angle medical images and to communicate with external receivers with higher energy efficiency in a small capsule size [2]. The typical frequency bands for the wireless capsule endoscopes are 401–406 MHz for the medical implant communication service (MICS) and MedRadio, and 433 MHz for the industrial-scientific-medical (ISM) services. As the capsule size is electrically very small compared with the frequency band, the antenna design significantly affects the form factor, wireless communication quality, and energy efficiency of the capsule. In addition, the impedance bandwidth of a capsule antenna should be wide. When the capsule travels through various tissue environments with different dielectric properties, it usually experiences an unstable contact condition in the GI tract. Then, the capsule antenna experiences severe impedance changes, which may lead to significant changes in the antenna characteristics. An ultra-wide bandwidth can mitigate such detuning effects.

In this work, we propose a conformal meandered loop antenna with a very wide bandwidth for sufficiently covering the MICS, MedRadio, and ISM bands. The omnidirectional radiation pattern of the antenna is verified by simulations and measurements, and its maximum gain is ~31.5 dBi. The fabricated antenna is successfully tested in an over-the-air wireless communication link, proving that the antenna can be instrumental in wireless capsule endoscopy applications.

Key Words: Capsule Endoscopy, Conformal Antenna, Meandered Loop Antenna, MICS, Wideband.
ed antenna is successfully verified in a 400-MHz over-the-air communication link system and proved to meet the requirements for capsule endoscopy applications.

II. ANTENNA DESIGN

One possible design approach for the capsule antenna is to embed an antenna inside a capsule [3–5]. However, the embedded type is usually not desired because the antenna dimension is inevitably limited and confined by the internal space of the capsule, which easily leads to the antenna performance degradation. Moreover, the internal antenna reduces the remaining inner space that is used by the battery and the electronic and optical components. This problem can be avoided by employing a conformal-type antenna. Previous conformal antennas for capsule endoscopes were based on a microstrip patch type [6–9], meandered line type, [10–13] and meandered loop type [14–17].

The patch type is usually a narrow band (only a small %), and always requires a solid ground plane that can block the camera view angle [6, 7]. A slight enhancement of the bandwidth was reported in [8, 9], but the technique was only implemented at a much higher frequency and not in the frequency of interest of this work (i.e., MICS, MedRadio, and ISM bands). The meandered line antennas showed a wider bandwidth and a smaller dimension. However, the meandered line antennas in [10, 11] were also realized in a higher frequency region.

Achieving a wide impedance bandwidth is important for capsule endoscopes because the dielectric property of human tissues is highly dispersive, showing significant changes with respect to different tissues and frequencies [14]. To cope with the highly dispersive property, a magnetic antenna in a loop structure is known to be less sensitive to the permittivity change. Thus, the loop antennas operating in the 400 MHz band were found to be effective, as they demonstrated a relatively wide bandwidth [15–17]. Accordingly, in this work, we have chosen the loop antenna structure.

Fig. 1 shows the proposed loop antenna structure. Many meanders are employed to realize a large electrical length while fitting to the outer dimension of the capsule. Fig. 1(a) shows the planar view. As the substrate is flexible, the planar structure can be wrapped in a cylindrical form, the front and back views of which are shown in Fig. 1(b) and (c). Fig. 1(d) presents the antenna dimensions. As shown in Fig. 1(b) and (c), the height and the diameter are 14 mm and 10 mm, which correspond to 0.152λ and 0.109λ respectively, with respect to the in-body wavelength λ of 91.7 mm. Note that the effective wavelength inside the human body is significantly smaller than that in free space because of the high permittivity (about 56.8) of the human body tissue, which has a dielectric loading effect [15]. The antenna is designed and fabricated in a polytetrafluoroethylene (PTFE) substrate with a thickness of 254 μm, relative permittivity εr of 2.17, and loss tangent tanδ of 0.0009. The copper thickness is 35 μm. The substrate is sufficiently flexible, and thus it can be bent and wrapped in a cylindrical form.

As the capsule travels in the GI tract, such as in the esophagus, stomach, colon, and small intestine, the capsule antenna experiences different human tissues with different dielectric properties and highly dispersive characteristics [14, 18]. The resonance characteristics of the proposed antenna are examined in various human tissue models, such as the stomach (εr = 67.19 and tanδ = 0.62), colon (εr = 62.01 and tanδ = 0.58), small intestine (εr = 65.2 and tanδ = 1.22), and human phantom (εr = 56.87 and tanδ = 0.59). The antenna is assumed to be placed at the center of a cylindrical human tissue model with a height of 120 mm and a diameter of 120 mm. Fig. 2 shows the simulated return losses in the four different tissue conditions. As can be seen, the proposed antenna exhibits a very wide bandwidth over the different tissue environments. The impedance bandwidth
with $S_{11}$ of less than $-10$ dB is 200 MHz–2 GHz, which greatly ensures the robust and stable antenna characteristics regardless of the tissue conditions. The wide bandwidth is accounted for by many parasitic capacitances introduced by the many closely spaced meandered conductor patterns. Similar effects can also be observed in previous antenna structures that have shown similar complex meandered patterns [9, 13, 15].

### III. MEASUREMENT RESULTS

The proposed antenna is fabricated. Its planar and cylindrical views are shown in Fig. 3(a) and (b), respectively. The antenna is measured with a differential feeding using a coaxial cable, and a balun is not used. As shown in Fig. 4, the return loss and radiation patterns of the fabricated antenna are measured after the antenna is placed at the center of a cylindrical bottle of liquid human phantom. Note that the liquid phantom is widely used in mimicking the human body [9, 13, 15–17], although minced pork can also be used [9, 12]. The height and the diameter of the cylindrical bottle are 190 mm and 50 mm, respectively. Adding salt and sugar to distilled water is known to decrease its relative permittivity while increasing its conductivity. The liquid human phantom in this work is made by a mixture of 51.8% distilled water, 46.75% sugar, and 1.45% salt, which gives $\varepsilon_r = 56.87$ and $\tan\delta = 0.59$ at 433 MHz [17].

Fig. 5 shows the measured return loss in comparison with the simulation result, which exhibits an excellent agreement. The measured bandwidth is 200 MHz–2.05 GHz, which corresponds to 164% of the fractional bandwidth. Note that such an ultra-wide bandwidth is highly desirable for the wireless capsule endoscope system. The wide bandwidth should result in a stable impedance matching and a robust operation for the capsule endoscope traveling through the GI tract and in a good coverage of the MICS band (402–405 MHz), the MedRadio band (401–406 MHz), and the ISM band (433.05–434.79 MHz and 902–928 MHz).

The radiation pattern is measured with the fabricated antenna placed in a closed bottle of liquid phantom. The radiation patterns are measured at 403 MHz. Fig. 6(a)–(c) show three different patterns at xy-cut, xz-cut, and yz-cut planes. Each plane with respect to the capsule orientation is depicted by the capsule image insets provided in each figure. As can be seen, the simulated and measured results show a reasonably good agreement. When the capsule travels through the GI tract, the orientation of the capsule is unknown and continuously changes. Therefore, to maintain a stable wireless link regardless of the capsule orientation and posture, the radiation pattern of a capsule antenna is preferred to be isotropic or at least omnidirectional [3, 5, 15]. As shown in Fig. 6, the radiation pattern of

![Fig. 2. Simulated return loss ($S_{11}$) of the proposed antenna in various tissues.](image)

![Fig. 3. Fabricated antenna in a planar (a) and a cylindrical form (b).](image)

![Fig. 4. Measurement setup with the fabricated antenna placed in a liquid human phantom.](image)

![Fig. 5. Simulated and measured return loss ($S_{11}$) of the fabricated antenna.](image)
This antenna is omnidirectional. The maximum gain is found to be –31.5 dBi at 403 MHz.

Over-the-air wireless communication link tests are performed by employing the fabricated antenna. Fig. 7 shows the wireless communication link setup. Commercially available two RF transceiver modules are used as the transmitter and the receiver. The fabricated antenna is used at the transmitter, and a commercial dipole antenna is used at the receiver. The capsule antenna is placed at the center of the liquid human phantom, and the receiver is placed 50 cm away. The number of transmitted packet per a single link is 10,000, and the data rate is 1.2 kbps. The transmit power is set to 0 dBm. Fig. 8 illustrates the measured received signal strength when the capsule-to-receiver distance changes from 0.1 m to 2 m. A theoretically computed curve considering the free space path loss is given for comparison with the measured data. As can be seen, the theoretical and measured data show a good agreement.

Table 1 summarizes the antenna performances and compares them with those in previous works. The impedance bandwidth of the proposed antenna significantly exceeds those in previous works. This antenna also achieves the miniature form factor and a reasonable radiation gain.

IV. CONCLUSION

We have presented a conformal ultra-wideband meandered loop antenna for wireless capsule endoscopy applications. The antenna dimensions are 14 mm in height and 10 mm in diameter. The antenna has a very wide bandwidth of 200 MHz–2.05 GHz, which corresponds to 164% of the fractional bandwidth. It shows an omnidirectional radiation pattern, and its measured maximum gain is –31.5 dBi. The over-the-air wireless communication link tests employing the fabricated antenna are successfully performed, proving the effectiveness of the proposed antenna for capsule endoscopy applications.
Table 1. Performance summary and comparison

<table>
<thead>
<tr>
<th>Antenna structure</th>
<th>This work</th>
<th>[17]</th>
<th>[16]</th>
<th>[15]</th>
<th>[13]</th>
<th>[12]</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operating frequency (MHz)</td>
<td>Meandered loop</td>
<td>433</td>
<td>433</td>
<td>401–406</td>
<td>433</td>
<td>433</td>
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<tr>
<td>Meandered loop</td>
<td></td>
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<tr>
<td>Meandered loop</td>
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<tr>
<td>Meandered wire</td>
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<tr>
<td>Meandered wire</td>
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<td></td>
</tr>
<tr>
<td>Impedance bandwidth(^a) (MHz)</td>
<td>200–2,050</td>
<td>336–1,080</td>
<td>327–530</td>
<td>370–630</td>
<td>284–825</td>
<td>371–496</td>
</tr>
<tr>
<td>Fractional bandwidth(^b) (%)</td>
<td>164</td>
<td>105</td>
<td>49</td>
<td>52</td>
<td>97</td>
<td>29</td>
</tr>
<tr>
<td>Dimension, h × d (mm)</td>
<td>14 × 10</td>
<td>20 × 11</td>
<td>17 × 7</td>
<td>17 × 11</td>
<td>15 × 6</td>
<td>21 × 12</td>
</tr>
<tr>
<td>Antenna gain (dBi)</td>
<td>-31.5</td>
<td>-28</td>
<td>-28.4</td>
<td>Not reported</td>
<td>-31.5</td>
<td>-36.9</td>
</tr>
<tr>
<td>Substrate (dielectric constant)</td>
<td>PTFE (2.17)</td>
<td>PPE plastic (2.55)</td>
<td>Polyimide (3.15)</td>
<td>Polyimide (3.15)</td>
<td>PTFE (2.55)</td>
<td>Polyimide (3.5)</td>
</tr>
</tbody>
</table>

\(^a\) Impedance bandwidth with -10 dB reflection coefficient.
\(^b\) Calculated by \((f_{\text{max}} - f_{\text{min}}) / f_{\text{center}}\).

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REFERENCES


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