I. INTRODUCTION

Recently, interest in distance estimation and direction-finding technology using ultra-wideband (UWB) has been increasing in various fields, such as the Internet of Things (IoT), vehicles, and smartphones [1–7]. A UWB system typically uses one transmit antenna and two or more receive antennas to estimate the direction-of-arrival (DoA) in the azimuth and elevation angles through phase difference [8, 9]. Therefore, when a UWB system is employed on a portable device, at least three antennas should be mounted to estimate the DoA in various directions. However, as portable devices are miniaturized, the size of the UWB array system must also be miniaturized [10]. Some studies have reduced the size of the antenna element [11, 12], such as corrugated edge structures [13], slotted radiators [4, 14, 15], and slot utilization on the ground plane [16, 17]. Although the antenna is miniaturized using these techniques, a relatively large space is still required to mount the array system. Recently, single radiator antennas were proposed to reduce the mounting space and simplify the array system [12, 18, 19]. However, these antennas were designed for ISM band, 5G MIMO communications, or L-band applications. Therefore, direction-finding UWB antennas that can be mounted on smart devices must still be investigated.

In this paper, we propose a UWB antenna using a single circular radiator with multiple ports (SRMP) to estimate the DoA. The proposed SRMP antenna is designed with multiple feeding ports in a single circular patch to minimize the UWB system by reducing the radiating elements. The UWB is a communication technology that utilizes the frequency spectrum between 3.1 GHz and 10.6 GHz [20]. The UWB frequency band is suitable for short-range high-data-rate communications due to its wide bandwidth and short pulse duration. Therefore, UWB technology is widely used in a variety of applications, including indoor localization, medical imaging, and machine-to-machine (M2M) communications. In this paper, we focus on the design of a UWB antenna with multiple ports on a single circular radiator for direction-finding applications.
method that uses the wideband of above the 500 MHz frequency band as defined by the Federal Communications Commission [20]. In general, the UWB band is defined from 3.1 GHz to 10.6 GHz according to 14 channel bands separated by the 500 MHz bandwidth [21]. In the Republic of Korea, only band 9 (7.737–8.236 GHz) is permitted as a UWB communication bandwidth for mobile devices. Therefore, the single circular patch is optimized on the FR-4 substrate to operate at 8 GHz, which corresponds to band 9 of UWB communications. Then, the three feeding ports are circularly arranged at intervals of 120° to estimate the DoA in the azimuth and elevation directions. To verify the feasibility, the proposed SRMP antenna is fabricated, and the antenna characteristics, such as the reflection coefficients, mutual couplings, and radiation patterns, are measured in a full anechoic chamber. To observe the direction-finding performance, the DoA estimation results using the Bartlett beamformer and the root-mean-square (RMS) error according to the signal-to-noise ratio (SNR) are obtained in the xy-plane. We then compare the DoA estimation performance of the proposed antenna with the typical circular array. The results demonstrate that the proposed antenna has an advantage in DoA estimation and is suitable for UWB direction-finding applications.

II. GEOMETRY AND PERFORMANCE OF THE PROPOSED ANTENNA

Fig. 1 presents the geometry of the proposed UWB antenna using a single circular patch radiator with multiple ports. The proposed SRMP antenna is designed considering the limited mounting space of smart devices by using a single circular patch that can fit into a small UWB system. Thus, the UWB antenna is designed on an FR-4 substrate with a diameter of \( l \) (0.53λ at 8 GHz) and a height of \( h \) (0.04λ at 8 GHz). A circular patch radiator with a diameter of \( d \) operating at 8 GHz is printed on the substrate, as shown in Fig. 1(a). Then, the three ports are arranged in a circle with intervals of \( \phi_0 \) at a distance of \( r_0 \) from the antenna center. The proposed SRMP antenna is modeled and simulated using the FEKO EM simulator, and the detailed optimized parameters are listed in Table 1. Photographs of the fabricated SRMP antenna, which is printed on the circular FR-4 substrate, are shown in Fig. 2. The proposed antenna is fed by three RG-402 cables that have low loss characteristics in the UWB frequency bands, as shown in Fig. 2(b).

The comparisons of the reflection coefficients and mutual couplings are shown in Fig. 3. The operating frequency band observed at port 1 has a bandwidth of 600 MHz (7.6–8.2 GHz) in the measurement and 500 MHz in the simulation. The proposed SRMP antenna is designed for UWB band 9, which requires a bandwidth of at least 500 MHz, because only band 9 (7.737–8.236 GHz) is permitted as a UWB communication bandwidth in Korea. The simulation of \(-13.3 \text{ dB}\) agrees well with the measurement of \(-13.3 \text{ dB}\) at 8 GHz. Mutual couplings are obtained between port 1 and port 2. The measured and simulated mutual couplings are \(-8.9 \text{ dB}\) and \(-8.8 \text{ dB}\) at 8 GHz, respectively. Considering the close port spacing of 4.5 mm (0.12λ), these coupling values can be regarded as relatively low values for small array systems. Of course, it is clear that an

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Value</th>
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<tbody>
<tr>
<td>( l )</td>
<td>20 mm</td>
</tr>
<tr>
<td>( h )</td>
<td>1.6 mm</td>
</tr>
<tr>
<td>( d )</td>
<td>10.9 mm</td>
</tr>
<tr>
<td>( r_0 )</td>
<td>2.5 mm</td>
</tr>
<tr>
<td>( \phi_0 )</td>
<td>120°</td>
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</table>

Fig. 1. Geometry of the proposed antenna: (a) isometric view and (b) bottom view.

Fig. 2. Photographs of the fabricated antenna: (a) top view and (b) side view.
array system using multiple radiators with array distance has better isolation characteristics than a single radiator antenna. However, it is difficult to apply it to mobile smart devices, where the physical size of the antenna should be very small.

Therefore, for the first time, we propose an SRMP antenna with the 600 MHz bandwidth as the second-best choice despite the low isolation characteristics.

The measured and simulated 2D radiation patterns of the proposed antenna in the zx- and zy-planes at 8 GHz for port 1 are shown in Fig. 4. The measured and simulated boresight gains are 3.4 dBi and 3.2 dBi, respectively. These results demonstrate that the proposed SRMP antenna is suitable for use in UWB applications despite the couplings due to the narrow port spacing.

III. DOA ESTIMATION USING THE BARTLETT BEAMFORMER

To examine the DoA estimation performance of the proposed antenna, the Bartlett beamformer, which is an extension of classical Fourier-based spectral analysis for array systems, is utilized. First, the received signal $x(t)$ in each antenna port is defined in Eq. (1). Vector $x(t)$ is the signals acquired at each port by the steering vector $a(\phi)$ of the array antenna, where an incident signal $s(t)$ is a continuous sine wave:

$$x(t) = a(\phi)s(t) + n(t).$$  

Vector $n(t)$ is additive white Gaussian noise for each port. The measured and simulated E-field ($E\_E$) radiation of the proposed antenna is shown in Fig. 5(a). It will be used as the steering vectors. To compare the DoA estimation performance of the proposed SRMP antenna with the typical array, we also simulate the E-field ($E\_E$) of a circular array with three antennas, as shown in Fig. 5(b). As can be seen, the maximum radiation of each pattern is oriented toward the direction in which each feeding port is placed. Then, the Bartlett beamformer denoted by Eq. (2) and Eq. (3) is used to examine the direction-finding performance. Vector $R$ is a covariance matrix of signal vectors that carries the spatial and spectral information of the incident signals, as expressed in Eq. (4):

$$P_{\text{mea}}(\phi) = \frac{a_{\text{mea}}(\phi)R_{\text{mea}}(\phi)a_{\text{mea}}^H(\phi)}{|a_{\text{mea}}(\phi)|^2},$$  

$$P_{\text{sim}}(\phi) = \frac{a_{\text{sim}}(\phi)R_{\text{sim}}(\phi)a_{\text{sim}}^H(\phi)}{|a_{\text{sim}}(\phi)|^2},$$  

$$R = x(t)x^H(t).$$

The power spectrum $P_{\text{mea}}$ can be derived with the steering vector $a_{\text{mea}}$ obtained by the measured radiation pattern. $P_{\text{sim}}$ is derived with the steering vectors $a_{\text{sim}}$ obtained by the simulated radiation pattern. This algorithm provides the maximum power spectrum at a given incident angle.

The power spectrum of the Bartlett beamformer in terms of the different incident angles are shown in Fig. 6. It is assumed that the SNR is 10 dB. In the first case, the signal is incident at $\phi_1 = 30^\circ$ on the xy-plane, and in the second case, it is $\phi_2 =$...
A 120°. The results computed with the measured and simulated data agree well (φ = 30° and φ = 29°, φ = 118° and φ = 117° for each incident angle), as shown in Fig. 6(a). The proposed antenna has an advantage in DoA estimation because the directivity of each pattern can improve the resolution of the power spectrum. The typical circular array with three antennas also has similar estimated angle results (φ = 29° and φ = 119° for each incident angle). However, the peak of the power spectrum is broad, as shown in Fig. 6(b), which leads to a lower resolution compared with the proposed SRMP antenna. In addition, we investigated the DoA estimation results in the elevation direction (in the zy-plane) to confirm that the proposed antenna is capable of direction-finding in the azimuth and elevation directions. The power spectrums of the Bartlett beamformer in the zy-plane with two different incident angles of θ₁ = 0° and θ₂ = 30°, as shown in Fig. 7. The DoA estimated by the measured radiation patterns has θ = 0° and θ = 31° as indicated by the solid and dotted lines, respectively. As expected, the results in the zy-plane show a more accurate DoA estimation compared to the results in the xy-plane.

Fig. 8 shows the RMS error of the DoA estimation as a function of the SNR. The RMS error is calculated by the
RMS difference between the actual and estimated signal directions for all incident angles in the \(xy\)-plane. To obtain the results, the average RMS error is computed with 100 iterations with random noise characteristics [18]. The RMS errors obtained using \(P_{\text{mea}}\) and \(P_{\text{sim}}\) are indicated by the solid and dashed lines. As expected, the RMS error decreases as the SNR increases. The proposed SRMP antenna and the circular array have an RMS error of less than 1° when the SNR is higher than 6 dB. The results verify that the proposed SRMP antenna is suitable for UWB direction-finding applications, despite using only one patch radiator.

IV. CONCLUSION

In this paper, we proposed a UWB SRMP antenna that can estimate the DoA using a single circular patch radiator. The single circular patch was optimized to operate at 8 GHz. Then, the three feeding ports were circularly arranged to achieve the DoA estimation in the azimuth and elevation directions. The proposed antenna simulated and measured reflection coefficients of \(-12.1\) dB and \(-13.3\) dB with a measured boresight gain of 3.4 dBi at 8 GHz. To observe the direction-finding performance, the DoA estimation results were obtained at incident angles of 30° and 120°. The power spectrums calculated from the measured and simulated patterns agreed well with each other as 29° and 30° and 118° and 117° for each incident angle, respectively. In addition, the RMS error in every incident angle was found to be less than 1° when the SNR was higher than 6 dB. The results verified that the proposed SRMP antenna was suitable for UWB direction-finding applications, despite using only one patch radiator.

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